

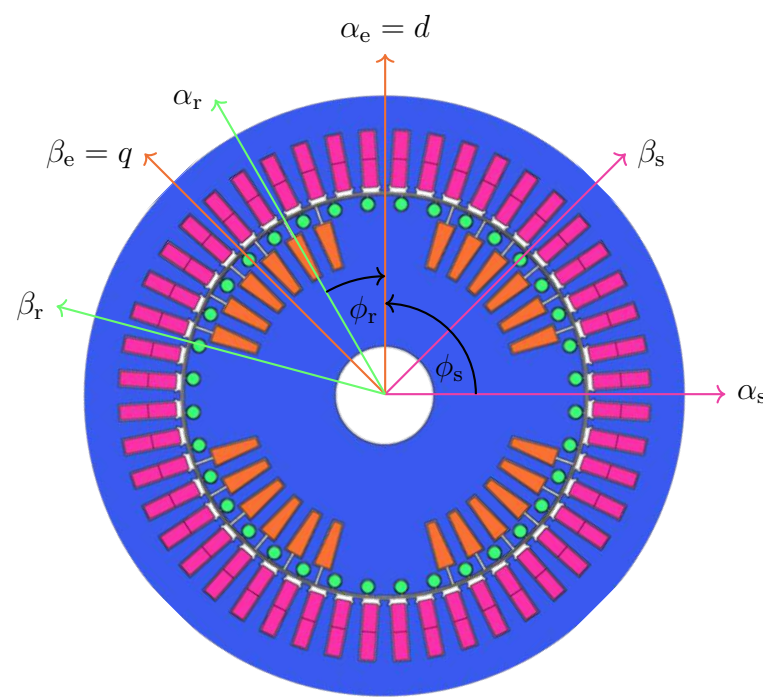
Derivation of Clarke and mutual inductance coupling factors in (d, q) -reference frame for synchronous machines with several multi-phase winding systems

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Motivation and problem statement

Example of a synchronous machine with several multi-phase winding systems showing steel sheets [■], stator winding [■], exciter winding [■], and damper winding [■].



How can we derive the Clarke factor and the mutual inductance coupling factors in the (d, q) -reference frame for such machines, considering saturation and cross-coupling effects?

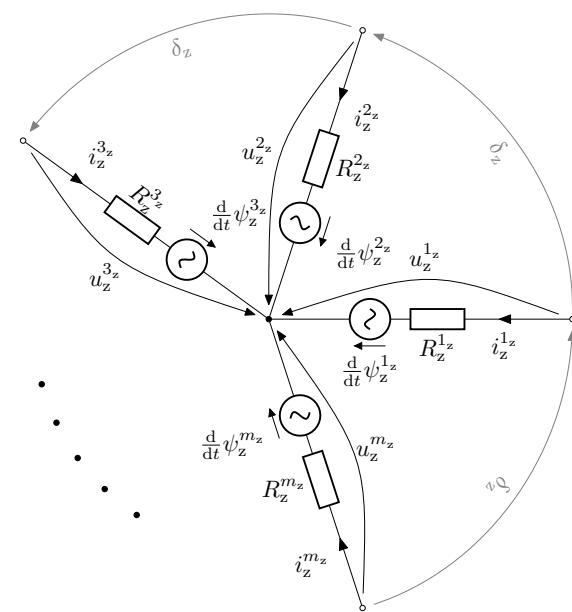
Generic modeling

Phase system with m_z phases

A multi-phase system with m_z phases is modeled by

$$\frac{d}{dt} \begin{pmatrix} \psi_z^{1z} \\ \vdots \\ \psi_z^{m_z} \end{pmatrix} = \begin{pmatrix} u_z^{1z} \\ \vdots \\ u_z^{m_z} \end{pmatrix} - \begin{pmatrix} R_z^{1z} & \cdots & 0 \\ \vdots & \ddots & \vdots \\ 0 & \cdots & R_z^{m_z} \end{pmatrix} \begin{pmatrix} i_z^{1z} \\ \vdots \\ i_z^{m_z} \end{pmatrix}. \quad (1)$$

$=: \psi_z^{1z \dots m_z} =: u_z^{1z \dots m_z} =: R_z^{1z \dots m_z} =: i_z^{1z \dots m_z}$



Clarke transformation

By applying the Clarke transformation matrix and its pseudo-inverse

$$\mathbf{T}_{c,z} := \kappa_z \begin{bmatrix} \cos(\phi_{1,z}) & \cdots & \cos(\phi_{m_z,z}) \\ \sin(\phi_{1,z}) & \cdots & \sin(\phi_{m_z,z}) \end{bmatrix} \iff \mathbf{T}_{c,z}^{-1} := \begin{bmatrix} \cos(\phi_{1,z}) & \sin(\phi_{1,z}) \\ \vdots & \vdots \\ \cos(\phi_{m_z,z}) & \sin(\phi_{m_z,z}) \end{bmatrix} = \tilde{\mathbf{T}}_{c,z}^T, \quad (2)$$

the orthogonal (α, β) -reference frame is obtained, leading to

$$\frac{d}{dt} \psi_z^{\alpha_z \beta_z} = u_z^{\alpha_z \beta_z} - R_z^{\alpha_z \beta_z} i_z^{\alpha_z \beta_z}. \quad (3)$$

Derivation of the Clarke factor

Amplitude invariant transformation is if the amplitude $\hat{x}$ of the original signal is equal to the norm of the transformed signal, i.e.

$$\hat{x} \stackrel{!}{=} \|\mathbf{x}_z^{\alpha_z \beta_z}\| = \|\mathbf{T}_{c,z} \mathbf{x}_z^{1z \dots m_z}\| = \left\| \kappa_z \tilde{\mathbf{T}}_{c,z} \hat{x} \underbrace{\begin{pmatrix} \cos(\phi_{1,z} + \phi_x(t)) \\ \vdots \\ \cos(\phi_{m_z,z} + \phi_x(t)) \end{pmatrix}}_{\text{symmetric signal}} \right\|. \quad (4)$$

The Clarke factor can be derived as

$$\kappa_z = \left\| \begin{bmatrix} \cos(\phi_{1,z}) & \cdots & \cos(\phi_{m_z,z}) \\ \sin(\phi_{1,z}) & \cdots & \sin(\phi_{m_z,z}) \end{bmatrix} \begin{pmatrix} \cos(\phi_{1,z} + \phi_x(t)) \\ \vdots \\ \cos(\phi_{m_z,z} + \phi_x(t)) \end{pmatrix} \right\|^{-1} = \begin{cases} \frac{1}{2}, & \text{if } m_z = 2 \text{ and } \phi_x(t) = 0, \\ \frac{2}{m_z}, & \text{if } m_z > 2. \end{cases} \quad (5)$$

Mutual inductance coupling factors

The partial differential equation of the flux linkages of the systems x and y is given by

$$\frac{d}{dt} \begin{pmatrix} \psi_x^{1x \dots m_x} \\ \psi_y^{1y \dots m_y} \end{pmatrix} = \begin{bmatrix} L_x^{1x \dots m_x} & M_{xy}^{1x \dots m_x} \\ M_{yx}^{1y \dots m_y} & L_y^{1y \dots m_y} \end{bmatrix} \frac{d}{dt} \begin{pmatrix} i_x^{1x \dots m_x} \\ i_y^{1y \dots m_y} \end{pmatrix} \quad (6)$$

$$\iff \frac{d}{dt} \begin{pmatrix} \psi_x^{\alpha_x \beta_x} \\ \psi_y^{\alpha_y \beta_y} \end{pmatrix} = \begin{bmatrix} L_x^{\alpha_x \beta_x} & M_{xy}^{\alpha_x \beta_x} \\ M_{yx}^{\alpha_y \beta_y} & L_y^{\alpha_y \beta_y} \end{bmatrix} \frac{d}{dt} \begin{pmatrix} i_x^{\alpha_x \beta_x} \\ i_y^{\alpha_y \beta_y} \end{pmatrix} \quad (7)$$

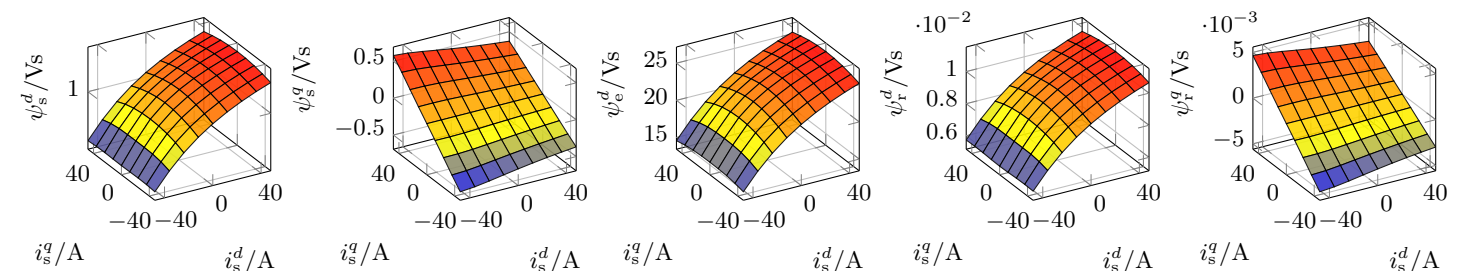
Due to the magnetic principle of reciprocity, $M_{xy}^{1x \dots m_x} = (M_{yx}^{1y \dots m_y})^T$ holds. However, after applying the Clarke transformation, the relationship between the mutual inductances of two arbitrary phase systems in orthogonal reference frames is characterized by the ratio $\frac{\kappa_x}{\kappa_y}$, i.e.

$$M_{xy}^{\alpha_x \beta_x} (M_{yx}^{\alpha_y \beta_y})^{-T} = \frac{\kappa_x}{\kappa_y} \underbrace{(\tilde{\mathbf{T}}_{c,x} M_{xy}^{1x \dots m_x} \tilde{\mathbf{T}}_{c,y}^T) (\tilde{\mathbf{T}}_{c,x} M_{yx}^{1y \dots m_y} \tilde{\mathbf{T}}_{c,y}^T)^{-1}}_{=I_2}. \quad (8)$$

Tailoring to EESM with damper winding

Nonlinear modeling and parameter identification by FEA

$$\left. \begin{aligned} \frac{d}{dt} \psi_s^{dq} &= u_s^{dq} - R_s^{dq} i_s^{dq} - \omega_p J \psi_s^{dq} = L_s^{dq} \frac{d}{dt} i_s^{dq} + L_{s,r}^{dq} \frac{d}{dt} i_r^{dq} + L_{s,e}^{dq} \frac{d}{dt} i_e^{dq}, \\ \frac{d}{dt} \psi_r^{dq} &= -R_r^{dq} i_r^{dq} + L_{r,e}^{dq} \frac{d}{dt} i_e^{dq} = L_{r,s}^{dq} \frac{d}{dt} i_s^{dq} + L_r^{dq} \frac{d}{dt} i_r^{dq} + L_{e,r}^{dq} \frac{d}{dt} i_e^{dq}, \\ \frac{d}{dt} \psi_e^d &= u_e^d - R_e^d i_e^d = L_{e,s}^d \frac{d}{dt} i_s^d + L_{e,r}^d \frac{d}{dt} i_r^d + L_e^d \frac{d}{dt} i_e^d, \\ \frac{d}{dt} \omega_m &= \frac{1}{\Theta_m} (m_m - m_l), \text{ and } \frac{d}{dt} \phi_m = \omega_m. \end{aligned} \right\} \quad (9)$$



FEA results confirm expected mutual inductance coupling factors:

$$L_{s,r}^{dq} = \frac{\kappa_s}{\kappa_r} (L_{r,s}^{dq})^T = 12 (L_{r,s}^{dq})^T, \quad L_{s,e}^{dq} = \frac{\kappa_s}{\kappa_e} (L_{e,s}^{dq})^T = \frac{4}{3} (L_{e,s}^{dq})^T, \quad L_{e,r}^{dq} = \frac{\kappa_e}{\kappa_r} (L_{r,e}^{dq})^T = 9 (L_{r,e}^{dq})^T.$$

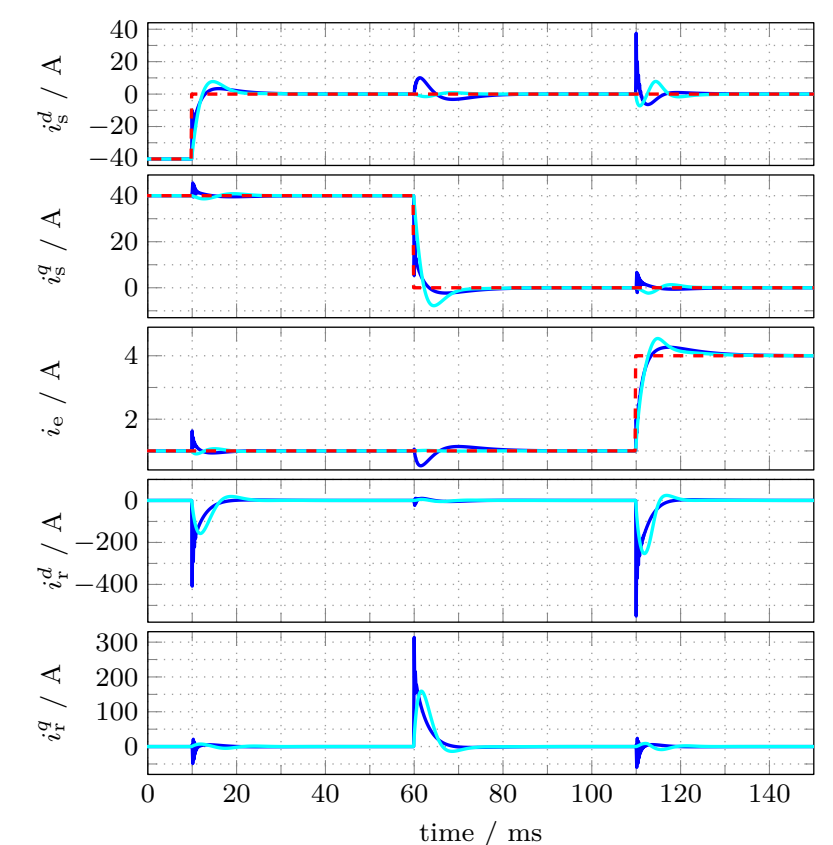
Simulation results: I/O-linearization based nonlinear current control

The right figure illustrates the time series results for reference current steps [---] within the current control system, involving the proposed model and identified parameters.

Two current controllers are compared:

(C₁) Controller considering the damper winding [—]

(C₂) Controller neglecting the damper effects [—]



References

- [1] Niklas Monzen, Johannes Rossmann, and Christoph M. Hackl. Derivation of clark and mutual inductance coupling factors in (d, q) -reference frame for synchronous machines with several multi-phase winding systems. In *Proceedings of the 2025 IECON Annual conference of the IEEE Industrial Electronics Society*, 2025.

