

DC-link control for airborne wind energy systems during pumping mode

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15.06.2015

TU Delft, Netherlands

Outline

- 1 Problem formulation
- 2 DC-link control
- 3 Conclusion and future work

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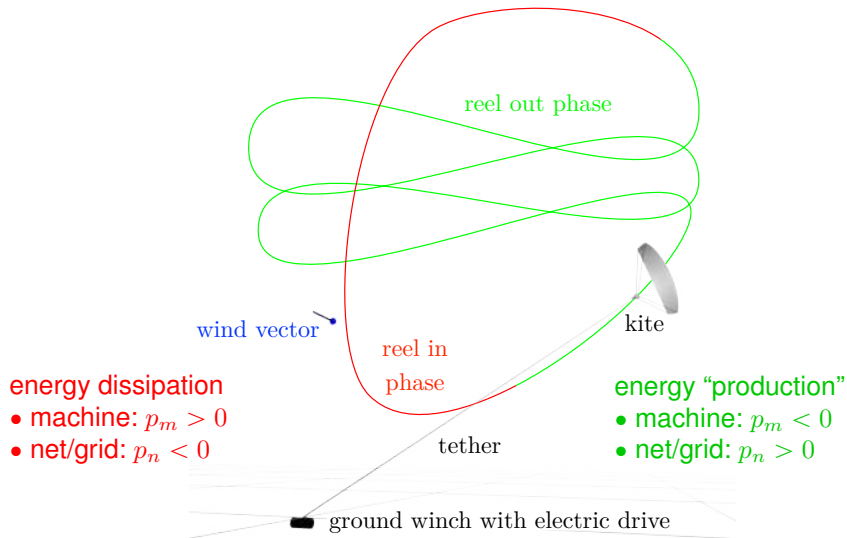
- 1 Problem formulation
 - Airborne wind energy systems during pumping mode
 - Overview and control system

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Problem formulation

Airborne wind energy systems during pumping mode $\Rightarrow$ **Bidirectional** power flow



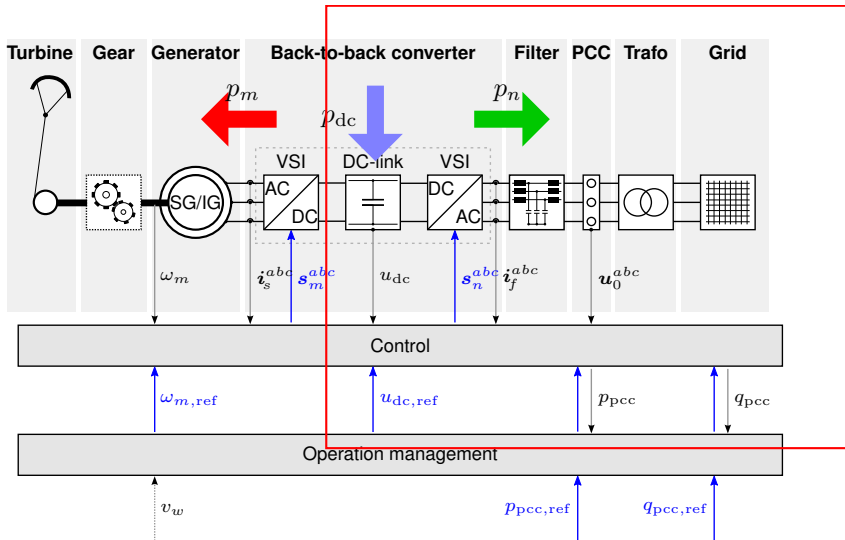
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Grid-side control

Overview



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2 DC-link control

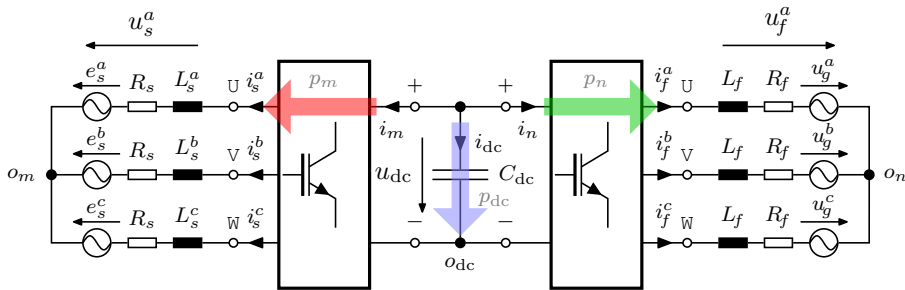
- Nonlinear dynamics of DC-link
- Classical DC-link PI controller
- Grid-side control structure in voltage orientation
 - Linearization and “structurally varying” system
 - “Structurally varying” system
 - Root locus and local stability analysis
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Nonlinear dynamics of DC-link



- $$p_{dc} = u_{dc} C_{dc} \frac{d}{dt} u_{dc} = -p_m - p_n \implies \frac{d}{dt} u_{dc} = \frac{1}{u_{dc} C_{dc}} (-p_m - p_n)$$
- $$p_m = \underbrace{\frac{3}{2} R_s \left\| \mathbf{i}_s^k \right\|^2}_{p_{m, \text{loss}}} + \underbrace{\frac{3}{2} p (\psi_{pm} i_s^q + (L_s^d - L_s^q) i_s^d i_s^q) \omega_m}_{p_{\text{mech}} = m_m \omega_m} + \underbrace{\frac{3}{2} \frac{d}{dt} (\mathbf{i}_s^k)^\top \mathbf{L}_s^k \mathbf{i}_s^k}_{p_{L_s^k}} =: p_m(t)$$
- $$p_n = \underbrace{\frac{3}{2} R_f \left\| \mathbf{i}_f^k \right\|^2}_{p_{pcc, \text{loss}}} + \underbrace{\frac{3}{2} \frac{d}{dt} L_f (\mathbf{i}_f^k)^\top \mathbf{i}_f^k}_{p_{L_f}} + \underbrace{\frac{3}{2} \hat{u}_g i_f^d}_{p_{pcc}}$$

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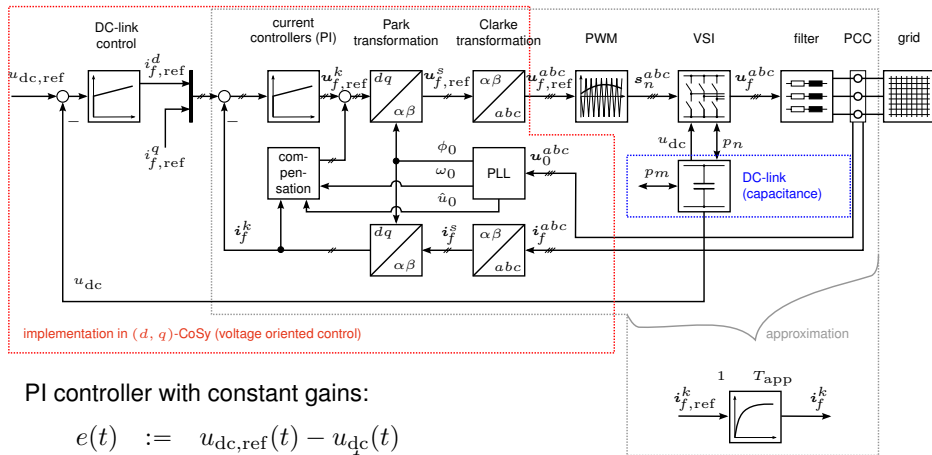
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Classical DC-link PI controller

Grid-side control structure in voltage orientation with **constant** PI controller



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Linearized system [1]

For small signals $\tilde{x} := x - x^*$, $\tilde{u} := u - u^*$, $\tilde{d} := d - d^*$, where

$$x := \begin{pmatrix} u_{dc} \\ i_f^k \end{pmatrix}, \quad u := i_{f,\text{ref}}^k = \begin{pmatrix} i_f^{d,*} \\ i_f^q \end{pmatrix}, \quad d := p_m \quad \text{and} \quad \begin{pmatrix} x^* \\ u^* \\ d^* \end{pmatrix} = \begin{pmatrix} u_{dc}^* \\ i_f^{k,*} \\ i_{f,\text{ref}}^k \\ p_m^* \end{pmatrix}, \quad (1)$$

it holds that

$$\frac{d}{dt}\tilde{x}(t) = A\tilde{x}(t) + B\tilde{u}(t) + b_d\tilde{d}(t)$$

with

$$A := \left. \frac{\partial f(x, u, d)}{\partial x} \right|_{(1)}, \quad B := \left. \frac{\partial f(x, u, d)}{\partial u} \right|_{(1)} = [b_1, b_2] \quad \text{and} \quad b_d := \left. \frac{\partial f(x, u, d)}{\partial d} \right|_{(1)}.$$

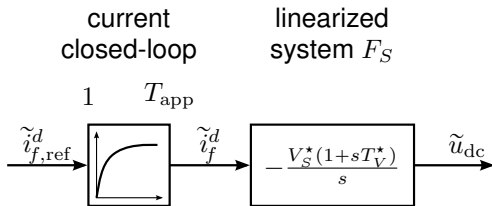
$$\begin{aligned} \Rightarrow \frac{\tilde{u}_{dc}(s)}{\tilde{i}_{f,\text{ref}}^d(s)} &= (1, 0, 0)(sI_3 - A)^{-1}b_1 = - \underbrace{\frac{3(\hat{u}_g + 2R_f i_f^{d,*})}{2C_{dc} u_{dc}^*}}_{=: V_S^* = V_S^*(x^*)} \frac{1 + s \underbrace{\left(\frac{L_f i_f^{d,*}}{\hat{u}_g + 2R_f i_f^{d,*}} \right)}_{=: T_V^* = T_V^*(x^*)}}{s(1 + sT_{\text{app}})} \\ &= - \frac{3(\hat{u}_g + 2R_f i_f^{d,*})}{2C_{dc} u_{dc}^*} \frac{1 + s \frac{L_f i_f^{d,*}}{\hat{u}_g + 2R_f i_f^{d,*}}}{s(1 + sT_{\text{app}})} \end{aligned}$$

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“Structurally varying” system [2]



$$\text{(Case 1)} \quad p_n = 0 \implies i_f^{d,*} = 0 \implies T_V^* = 0 \implies \frac{\tilde{u}_{dc}(s)}{\tilde{i}_{f,ref}^d(s)} = \frac{V_S^*}{s(1+sT_{app})}$$

$$\text{(Case 2)} \quad p_n > 0 \implies i_f^{d,*} > 0 \implies T_V^* > 0 \implies \frac{\tilde{u}_{dc}(s)}{\tilde{i}_{f,ref}^d(s)} = \frac{V_S^*(1+sT_V^*)}{s(1+sT_{app})}$$

(minimum-phase)

$$\text{(Case 3)} \quad p_n < 0 \implies i_f^{d,*} < 0 \implies T_V^* < 0 \implies \frac{\tilde{u}_{dc}(s)}{\tilde{i}_{f,ref}^d(s)} = \frac{V_S^*(1-s|T_V^*|)}{s(1+sT_{app})}$$

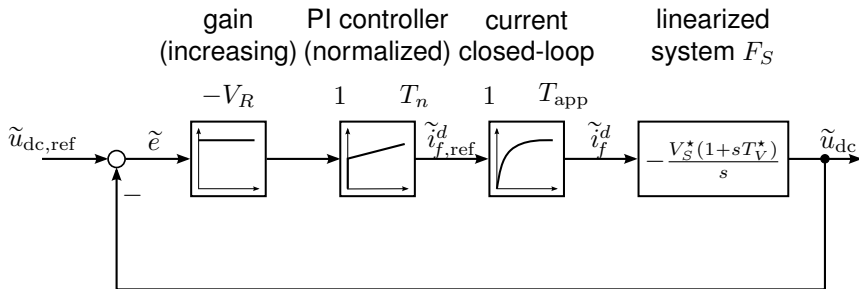
(nonminimum-phase)

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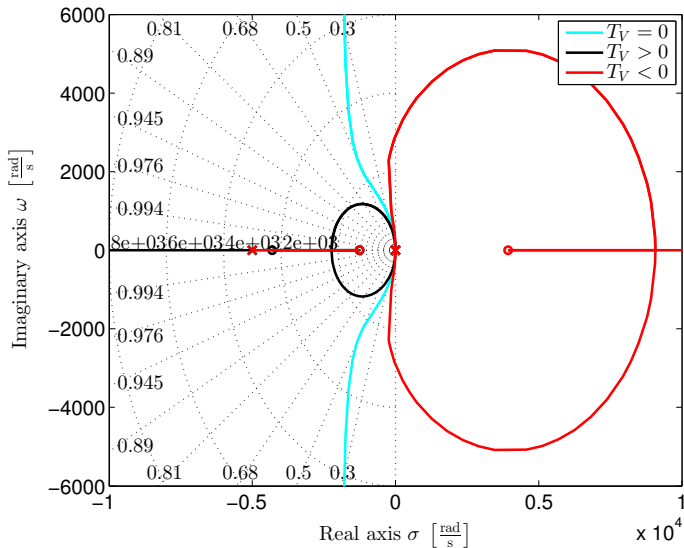
Root locus of linearized system



open loop of linearized system

$$\frac{\tilde{u}_{dc}(s)}{\tilde{e}(s)} = \frac{V_R V_S^* (1 + sT_n)(1 + sT_V^*)}{s^2 T_n (1 + sT_{app})}, \quad \begin{cases} T_{app}, T_n > 0, V_S^* > 0 \\ T_V^* \in \mathbb{R} \text{ (depends on equilibrium!)} \\ V_R \rightarrow \infty \end{cases}$$

Root locus ($T_V = T_V^*$ depends on equilibrium!)



Local stability analysis

Hurwitz criterion

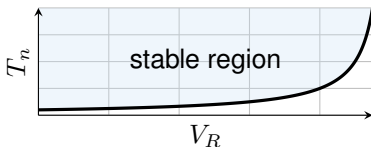
Transfer function of (linearized) closed-loop system

$$F_{CL}(s) = \frac{u_{dc}(s)}{u_{dc,ref}(s)} = \frac{V_R V_S^* (1 + sT_{app})(1 + sT_V^*)}{s^3 \underbrace{T_n T_{app}}_{=:a_3} + s^2 \underbrace{T_n (1 + T_V^* V_S^* V_R)}_{=:a_2} + s \underbrace{(T_V^* + T_n) V_S^* V_R}_{=:a_1} + \underbrace{V_R V_S^*}_{=:a_0}}$$

Sufficient condition for *local* stability

- $0 < V_R < \frac{1}{|T_V^*| V_S^*}$
- $T_n > \frac{T_{app}}{1 - V_R V_S^* |T_V^*|} + |T_V^*| > 0$

$$\implies a_0, a_1, a_2, a_3 > 0 \text{ and } D_2 := a_2 a_1 - a_3 a_0 > 0.$$



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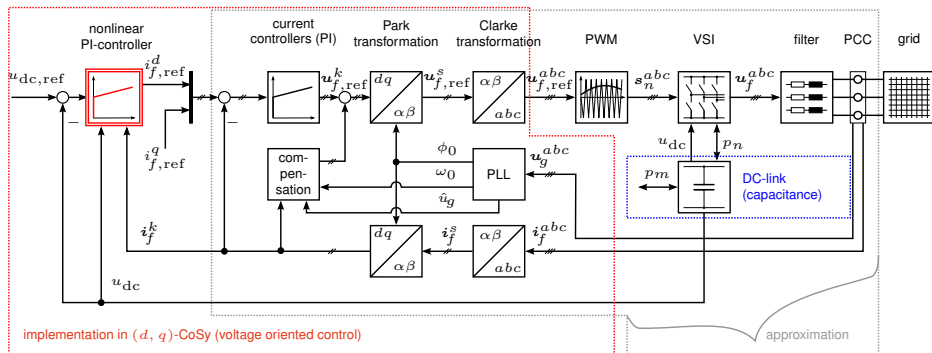
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Nonlinear DC-link PI controller

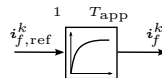
Control system in voltage orientation with **nonlinear** PI controller



Nonlinear PI controller

$$e(t) := u_{dc,ref}(t) - u_{dc}(t)$$

$$i_{f,ref}^d(t) = V_R(i_f^k, u_{dc}) e(t) + \frac{V_R(i_f^k, u_{dc})}{T_n(i_f^k, u_{dc})} \int_0^t e(\tau) d\tau$$



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Nonlinear DC-link PI controller

Controller design and local stability analysis

Pole placement: Two tuning parameters V_R & T_n ; three poles λ_1 & $\lambda_2 \pm j\lambda_i$

$$\underbrace{s^3 T_n T_{app}}_{=:a_3} + \underbrace{T_n (1 + T_V^* V_S^* V_R)}_{=:a_2} + s \underbrace{(T_V^* + T_n) V_S^* V_R}_{=:a_1} + \underbrace{V_R V_S^*}_{=:a_0} \\ \stackrel{!}{=} (s - \lambda_1)(s - \lambda_2 + j\lambda_i)(s - \lambda_2 - j\lambda_i)$$

For given $\lambda_2 < 0$ and $\lambda_i \neq 0 \implies$ Computation of $V_R(T_V^*, V_S^*, \lambda_2, \lambda_i) > 0$, $T_n(T_V^*, V_S^*, \lambda_2, \lambda_i) > 0$ (analytic expression derived for online parameter adjustment) and $\lambda_1(T_V^*, V_S^*, \lambda_2, \lambda_i) < 0$.

Sufficient condition for *local* stability

- $0 < \lambda_i^2 < \frac{1}{|T_V^*| T_{app}}$
- $0 > \lambda_2 \geq \max \left\{ -\frac{1}{2T_{app}}, \frac{1}{|T_V^*|} - \sqrt{\frac{1}{(T_V^*)^2} + \frac{1}{|T_V^*| T_{app}}} - \lambda_i^2 \right\}$

$\implies a_0, a_1, a_2, a_3 > 0, D_2 := a_2 a_1 - a_3 a_0 > 0$ and $\lambda_1 < 0$.

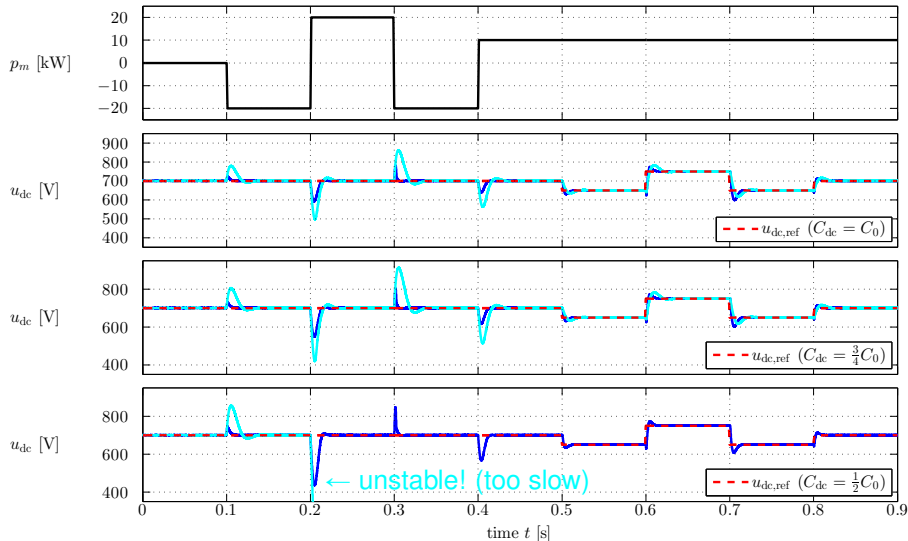
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Nonlinear PI-control for DC-link voltage

Simulation results for 20 kW AWE system: — constant PI, — nonlinear PI



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3 Conclusion and future work

Conclusion and future work

To take home:

- DC-link voltage control
 - nonlinear system dynamics with structurally varying nature
 - **non**minimum-phase behavior during reel-in phase
 - very conservative classical PI controller design necessary
- nonlinear PI controller
 - design based on pole placement and online parameter adjustment
 - faster tracking and disturbance rejection
 - smaller DC-link capacitance feasible (→ cheaper)

Future work:

- Lyapunov-based stability analysis
- Measurement results (construction of test bench in progress)

References

- [1] C. Dirscherl, C. Hackl, and K. Schechner, “Explicit model predictive control with disturbance observer for grid-connected voltage source power converters,” in *Proceedings of the 2015 IEEE International Conference on Industrial Technology*, pp. 999–1006, 2015.
- [2] C. Dirscherl, C. Hackl, and K. Schechner, “Modellierung und Regelung von modernen Windkraftanlagen: Eine Einführung (*available at the authors upon request*),” in *Elektrische Antriebe – Regelung von Antriebssystemen* (D. Schröder, ed.), ch. 24, Springer-Verlag, 2015.